### Last week

- State Space Realizations (pp 139-150)
- G(s), denominator and numerator, poles and zeros
- Change of coordinates, diagonal and controllable form
- State-feedback
- Observers
- Feedback from estimated states
- Integral action by disturbance model

## Lecture 5

- Controllability Existence of control signal
  - Which state directions can be controlled ?
- Observability Determine state
  - Which state directions can not be seen?
- Kalman's decomposition theorem
- Cancelled dynamics <=> lack of controllability or observability

# Controllability

How should controllability be defined ?

Some (not used) alternatives:

By proper choice of control signal u

- any state  $x_0$  can be made an equilibrium
- any state trajectory x(t) can be obtained
- any output trajectory y(t) can be obtained

The most fruitful definition has instead turned out to be the following

# Controllability

The state equation

$$\dot{x}(t) = Ax(t) + Bu(t), \quad x(0) = x_0$$

is called *controllable* if for any  $x_0$  and T > 0, there exists u(t) such that x(T) = 0 ("Controllable to origin")

Question: Is this equivalent to the following definition:

"for  $x_0 = 0$  and any  $x_1$  and T > 0, there exists u(t) such that  $x(T) = x_1$ " ("Controllable from origin")

The audience is thinking!

Hint: 
$$x(T) = e^{AT}x_0 + \int_0^T e^{A(T-t)}Bu(t)dt$$

### **Controllability Gramian**

The matrix

$$W(T) = \int_0^T e^{-At} B B^T e^{-A^T t} dt$$

is called the controllability Gramian.

Note that it is positive semidefinite,  $W(T) \ge 0$ 

The main controllability result is the following

### **Theorem Controllability Test**

The following conditions are equivalent:

- (i) The system  $\dot{x}(t) = Ax(t) + Bu(t)$  is controllable.
- (ii) rank  $[B \ AB \ A^2B \ \dots \ A^{n-1}B] = n.$
- (iii) W(T) is invertible for any T > 0
- (iv) For any  $\lambda \in \mathbf{C}$  we have rank $[A \lambda I \ B] = n$

We will prove  $(i) \Rightarrow (ii) \Rightarrow (iii) \Rightarrow (i)$ 

The condition (iv), not proved here, is called the PBH test (Popov-Belevitch-Hautus).

# Analysing controllability

The system is by definition controllable iff we for any  $x_0$  and T can find control signal  $u(t), t \in [0, T]$  that solves (see hint some slides above)

$$-x_0 = \int_0^T e^{-At} Bu(t) dt \tag{(\star)}$$

Cayley-Hamilton's theorem (google it) says that  $A^k$  for  $k \ge n$  can be written as a linear combination of  $I, A, A^2, \ldots A^{n-1}$ , so

$$e^{-At} = \sum_{k=0}^{\infty} \frac{(-t)^k}{k!} A^k = \sum_{k=0}^{n-1} f_k(t) A^k$$
, (for some  $f_k(t)$ ).

Therefore the condition (\*) can be written

$$-x_0 = [B \ AB \ A^2B \ \dots \ A^{n-1}B]F(u), \tag{**}$$

for some vector F(u) with elements  $F_k(u) = \int_0^T f_k(t)u(t)dt$ 

# **Proof of** $(i) \Rightarrow (ii)$

Proof by contradiction: Assume (ii) does not hold, i.e. the controllability matrix does not have full rank.

This means there is a vector, lets call it  $-x_0$ , that is not in the column span of

$$[B AB A^2B \dots A^{n-1}B]$$

This contradicts (\*\*), so (i) does not hold.

# **Proof of** $(ii) \Rightarrow (iii)$

Assume (iii) does not hold. Then there is a  $p \neq 0$  so W(T)p = 0.

$$0 = p^T W(T) p = \int_0^T \left( p^T e^{-At} B \right) \left( B^T e^{-A^T t} p \right) dt$$

Therefore

$$p^T e^{-At} B = 0, \forall t.$$

Derivating this k times and setting t = 0 gives  $p^T A^k B = 0$ . Hence we have

$$p^T[B AB A^2B \dots A^{n-1}B] = 0.$$

Therefore (ii) does not hold.

# $(iii) \Rightarrow (i)$ Explicit construction of u(t)

If W(T) is invertible, then for any initial state  $x_0$ , the control signal

$$u(t) = -B^T e^{-A^T t} (W(T))^{-1} x_0$$

gives x(T) = 0 (check that (\*) some slides before is satisfied!). Hence the system is controllable.

# Another interpretation of W(T)

One can prove (using techniques from next lecture) that the minimal (squared) control energy, defined by  $||u||^2 := \int_0^T |u|^2 dt$ , needed to move from  $x(0) = x_0$  to x(T) = 0 equals

 $x_0^T (W(T))^{-1} x_0$ 

Gives nice formula for which state directions are costly to control.

W(T) large in some direction means easy to control in that direction

# Which trailer is controllable?









# Observability

The system

$$\begin{cases} \frac{dx}{dt} = Ax, \quad x(0) = x_0\\ y = Cx \end{cases}$$

is called observable if  $x_0$  can be uniquely determined from  $y_{[0,T]}$  (for any  $T>0{\rm )}$ 

This is the same as saying that the only  $x_0$  for which y(t) = 0 for all t is the trival case  $x_0 = 0$ 

WHY ? The audience is thinking!

## Which trailer is observable?



### **Theorem - Observability Criteria**

The following are equivalent

(i) The system  $\dot{x} = Ax$ , y = Cx is observable (ii)  $\operatorname{rank} \begin{pmatrix} C \\ CA \\ \vdots \\ CA^{n-1} \end{pmatrix} = n$ (iii)  $\widetilde{W}(T)$  is invertible for any T > 0(iv) For any  $\lambda \in \mathbf{C}$  we have  $\operatorname{rank} \begin{pmatrix} A - \lambda I \\ C \end{pmatrix} = n$ 

Here the observability Gramian W(T) is defined as

$$\widetilde{W}(T) = \int_0^T e^{A^T t} C^T C e^{At} dt$$

# Proof that (i) ⇔ (ii)

If (i) does not hold, then there is a quiet state  $x_0 \neq 0$  so that

$$y(t) = Ce^{At}x_0 = 0, \quad \forall t$$

Derivating this k times and setting t = 0 we get  $CA^k x_0 = 0$ . This shows (ii) doesn't hold.

On the other hand, if (ii) does not hold then a nonzero  $x_0$  can be found so  $CA^k x_0 = 0$  for k = 0, ..., n - 1. By Cayley-Hamilton this means  $CA^k x_0 = 0$  also for  $k \ge n$ , so by power expansion of  $e^{At}$ 

$$y(t) = Ce^{At}x_0 = 0, \quad \forall t,$$

which says that (i) does not hold.

# Proof that (ii) $\Leftrightarrow$ (iii)

This follows easily by substituting (A,B) with  $(A^T,C^T)$  in (ii) and (iii) in the controllability theorem earlier

This illustrates a so called **duality** between the two theorems



# Bonus Proof of (iii) $\Rightarrow$ (i)

Maybe you didn't like the earlier proof of (ii)  $\Rightarrow$  (i) that used derivation of y(t). It is hard to implement in practice. If there e.g. is measurement noise on y(t) we would like a better way of determining  $x_0$ .

The dual result of the construction of the energy-optimal u(t) in the controllability theorem is the following calculation:

Since  $y(t) = Ce^{At}x_0$  we have that

$$\int_0^T e^{A^T t} C^T y(t) dt = \widetilde{W}(T) x_0$$

When  $\widetilde{W}(t)$  is invertible we can hence find  $x_0$  by

$$x_0 = (\widetilde{W}(T))^{-1} \int_0^T e^{A^T t} C^T y(t) dt.$$

This way of determining  $x_0$  is actually optimally robust against measurement noise, in a sense described in Lecture 6





### Controllability – state transformation

Theorem:

If the system is noncontrollable, say rank(C) = q < n, then there is a state transformation x = Vz so that in the new state coordinates

$$AV = V \begin{pmatrix} \tilde{A}_{11} & \tilde{A}_{12} \\ 0 & \tilde{A}_{22} \end{pmatrix}$$
 and  $B = V \begin{pmatrix} \tilde{B}_1 \\ 0 \end{pmatrix}$ ,

 $( ilde{A}_{11}, ilde{B}_1)$  controllable subsystem, q imes q

### **Observability – state transformation**

Theorem:

If the system is non-observable, say rank(O) = q < n, then there is a state transformation so that in the new state coordinates

$$AV = V \begin{pmatrix} A_{11} & 0\\ \tilde{A}_{21} & \tilde{A}_{22} \end{pmatrix}$$
 och  $CV = \begin{pmatrix} \tilde{C}_1 & 0 \end{pmatrix}$ ,

 $(\tilde{A}_{11},\tilde{C}_1)$  observable subsystem, q imes q

With a state transformation that splits the controllable subspace (and its complement) into nonobservable subspace and complement we get the system on a nice form

$$\frac{dx}{dt} = \begin{pmatrix} A_{11} & 0 & A_{13} & 0 \\ A_{21} & A_{22} & A_{23} & A_{24} \\ 0 & 0 & A_{33} & 0 \\ 0 & 0 & A_{43} & A_{44} \end{pmatrix} x + \begin{pmatrix} B_1 \\ B_2 \\ 0 \\ 0 \end{pmatrix} u$$
$$y = \begin{pmatrix} C_1 & 0 & C_2 & 0 \end{pmatrix} x$$

$$G(s) = C_1(sI - A_{11})^{-1}B_1$$

Illustrates what subparts of the system that influences the input-output behavior

### Kalman's decomposition theorem



The audience if thinking: What blocks in this figure corresponds to parts 1,2,3,4 on the previous slide?

# Kalman's decomposition theorem



If no common eigenvalues between any two blocks on the diagonal, then corresponding off-diagonal blocks can be eliminated by changed choice of the complementing spaces. Simplifies picture further





What does the decomposition theorem say when  $y = \theta_2$ ? What block is then missing?

## Trailer 4 after coordinate change

$$\begin{bmatrix} \dot{\theta}_2\\ \dot{\theta}_3\\ \dot{\theta}_1 - \dot{\theta}_2 \end{bmatrix} = \begin{bmatrix} -1 & 0 & 0\\ 2 & -2 & 0\\ 0 & 0 & -1 \end{bmatrix} \begin{bmatrix} \theta_2\\ \theta_3\\ \theta_1 - \theta_2 \end{bmatrix} + \begin{bmatrix} 1\\ 0\\ 0 \end{bmatrix} u$$
$$y = \begin{bmatrix} 1 & 0 & 0 \end{bmatrix} \begin{bmatrix} \theta_2\\ \theta_3\\ \theta_1 - \theta_2 \end{bmatrix}$$

controllable and observable subsystem:  $\theta_2$ 

### Zeros and state feedback

Remember: State-feedback does not change zeros. Choose state feedback L that gives a pole in  $\lambda$ . If the mode  $x_0 e^{\lambda t}$  now becomes non-observable

$$\begin{pmatrix} A - BL - \lambda I \\ C \end{pmatrix} x_0 = 0$$

then actually  $\lambda$  was a zero to the system:

$$\begin{pmatrix} A - \lambda I & B \\ C & 0 \end{pmatrix} \begin{pmatrix} x_0 \\ u_0 \end{pmatrix} = 0$$

Corresponds to cancellation of the factor  $s - \lambda$  in

$$G(s) = C(sI - A + BL)^{-1}Bl_r$$

#### **Bonus: Series Connection SISO**

Given two systems  $n_i(s)/d_i(s) = c_i(sI - A_i)^{-1}b_i$ , i = 1, 2Then the series connection  $\frac{n_2(s)}{d_2(s)}\frac{n_1(s)}{d_1(s)}$  is

- uncontrollable  $\iff$  there is  $\lambda$  so  $n_1(\lambda) = d_2(\lambda) = 0$
- unobservable  $\iff$  there is z so  $n_2(\lambda) = d_1(\lambda) = 0$

Proof:

Controllable, check when rank 
$$\begin{bmatrix} \lambda I - A_1 & 0 & b_1 \\ -b_2 c_1 & \lambda I - A_2 & 0 \end{bmatrix} \le n$$
  
Observable, check when rank 
$$\begin{bmatrix} \lambda I - A_1 & 0 \\ -b_2 c_1 & \lambda I - A_2 \\ 0 & c_2 \end{bmatrix} \le n$$

#### **Cancellation in series connections**

Example

$$Y(s) = \frac{s+3}{s-1} \cdot \frac{s-1}{s+2} U(s)$$

Loss of controllability of an unstable mode. Bad.

Example

$$Y(s) = \frac{s-1}{s+2} \cdot \frac{s+3}{s-1} U(s)$$

Loss of observability of an unstable mode. Also bad.

### Summary

- Controllability criteria
- Observability criteria
- Kalman's decomposition
- Cancelled dynamics <=> lack of controllability or observability